

TIME DOMAIN ANALYSIS (A VERY SUPERFICIAL APPROACH)





Laplace Transform (LT)

- The Laplace Transform (LT) is an integral transform similar to the Fourier Transform
- The LT of a function f(t) is defined as

$$F(s) = \mathcal{L}\left\{f
ight\}(s) = \int_0^\infty f(t)e^{-st}\,\mathrm{d}t,$$

- s is a *complex* variable
- This integral does not exist for all possible f(t) and s!
- (If s has a real part >0, f(t) must not grow faster than C e^{Re (s) t})
- The Inverse Transform is more complicated:

$$\mathcal{L}^{-1}\left\{F(s)
ight\} = rac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} F(s) \, ds$$

- where γ > Re(all singularities of f).
- This is a line integral in the complex s-plane, 'right' of all singularities

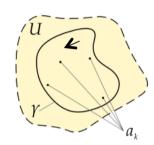




Reminder (hopefully..): Integration with Residues

The statements are valid under certain conditions only. Complex Analysis! ■ The *Residue Theorem* states that the line integral of a function f(z) along a closed curve γ in the complex z-plane is $2\pi i \times (\text{the sum of the residues at the } singularities} a_k \text{ of } f)$:

$$\oint_{\gamma} f(z) dz = 2\pi i \sum_{k} \operatorname{Res}(f, a_k)$$



Wikipedia

- The *residue* is a characteristic of a singularity a_k (or c below)
 - For a first order (simple) pole at c (f behaves ~ like 1/z at c):

$$Res(f,c) = \lim_{z \to c} (z - c) f(z).$$

• For a pole of order n:

Res
$$(f,c) = \frac{1}{(n-1)!} \lim_{z \to c} \frac{d^{n-1}}{dz^{n-1}} ((z-c)^n f(z)).$$





Example for Integration with Residues

- Assume we want to find $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1}$.

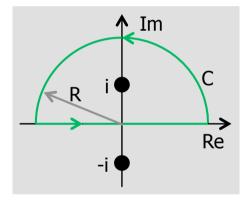
 The function $f(z) = \frac{1}{z^2 + 1} = \frac{1}{(z + i)(z i)}$ has poles i and -i

■ The residue at *i* is:

$$Res(f, i) = \lim_{z \to i} f(z)(z - i) = \lim_{z \to i} \frac{1}{(z + i)} = \frac{1}{2i}$$

■ The line integral along green curve C is

$$\int\limits_C f(z)dz = 2\,\pi\,i\,\operatorname{Res}(f,i) = \pi$$



- When we increase the size of the curve, the contribution of the upper arc vanishes* (the length of the arc rises ~R, but f falls as 1/R2)
- Therefore $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1} = \pi$





Example 1: LT of 1

• For f(t) = 1:

L {1} =
$$\int_0^\infty e^{-st} dt = \left(\frac{-e^{-st}}{s}\right)_0^\infty = \frac{-e^{-s\infty}}{s} + \frac{e^{-s0}}{s} = \frac{1}{s}$$

Valid for Re[s] > 0 to make sure this vanishes

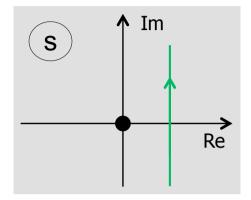




Example 1 (inverse): Inverse LT of 1/s

• For
$$F(s) = \frac{1}{s}$$
:

$$L^{-1}\left\{\frac{1}{s}\right\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} \frac{1}{s} ds$$



- The integral has just one pole at s = 0.
- The Residuum is:

Res
$$\left[e^{st}\frac{1}{s}, 0\right]$$
 = Limit $\left[(s-0)e^{st}\frac{1}{s}, s \rightarrow 0\right]$ = Limit $\left[e^{st}, s \rightarrow 0\right]$ = 1

So the Integral is

$$\int_{\gamma-i\infty}^{\gamma+i\infty} \dots ds = 2\pi i \operatorname{Res}[\dots, 0] = 2\pi i$$

■ And we just have
$$L^{-1}\left\{\frac{1}{s}\right\} = 1$$





Properties of Laplace Transforms

$$lacksquare ext{For } f(t) = \mathcal{L}^{-1}\{F(s)\}, \, g(t) = \mathcal{L}^{-1}\{G(s)\} \, ext{ we have:}$$

Function	Laplace Transform	
af(t)+bg(t)	aF(s)+bG(s)	Linearity
f'(t)	sF(s)-f(0)	Derivative
$(fst g)(t)=\int_0^t f(au)g(t- au)d au$	$F(s)\cdot G(s)$	Convolution
$\int_0^t f(\tau)d\tau = (u*f)(t)$	$\frac{1}{s}F(s)$	Integration
f(t-a)u(t-a)	$e^{-as}F(s)$	Time Shift

u(t): Step function



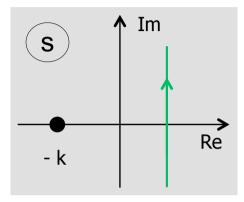


Example 2 ('frequency shift'):

• For F(s) =
$$\frac{1}{s + k}$$
:

$$L^{-1} \{F[s]\} = \frac{1}{2 \pi i} \int_{\gamma - i \infty}^{\gamma + i \infty} e^{st} \frac{1}{s + k} ds$$

■ The pole is now at s = -k.



The Residuum is:

Res
$$\left[e^{st} \frac{1}{s+k}, 0\right]$$
 = Limit $\left[\left(s+k\right) e^{st} \frac{1}{s+k}, s \rightarrow -k\right] = e^{-kt}$

■ And we just have
$$L^{-1}\left\{\frac{1}{s+k}\right\} = e^{-kt}$$





Why is Laplace Transform so Useful?

- Differential / Integral equations in t can be converted to Analytical equations in s, where they can be solved
- EQ(t) \rightarrow transform to H(s) \rightarrow Solve in s \rightarrow Transform back
- Example: Radioactive Decay
 - f[t]: Number of atoms at time t
 - The # of decaying atoms is prop. to # of atoms: $\frac{df[t]}{dt} = -\lambda f[t]$
 - With F[s] = LT(f[t]):
 (f[0] = N₀ is initial number of atoms)

$$sF[s] - f[0] = -\lambda F[s]$$

• This can be solved in s-domain:

$$F[s] = \frac{N_0}{s + \lambda}$$

Transforming back (see example) gives: f[t] = N_θ e^{-λt}



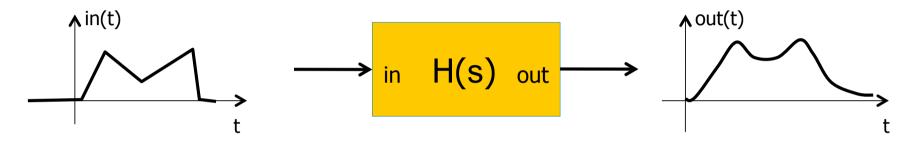
LAPLACE TRANSFORM AND TRANSFER FUNCTION





Time Response

- The **transfer function** tells us how sine inputs are modified by the system, i.e. what happen in the **frequency domain**
- How can be get the time response for an arbitrary input?

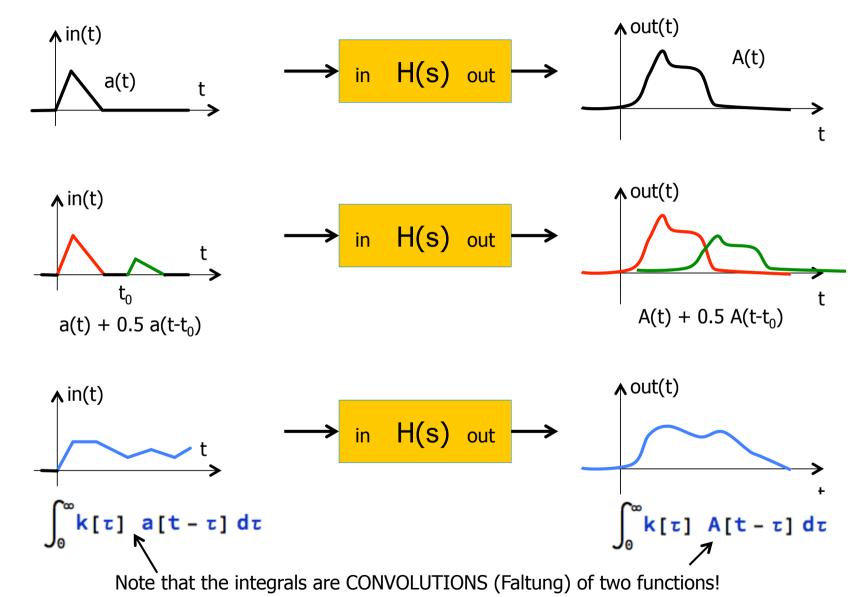


- For a *linear, time invariant (LTI)* system, we can use:
 - The response of a k × larger input pulse is just k × larger
 - The response for a time shifted input is time shifted
- For such a system we can
 - express the input signal as a superposition of 'simple' signals
 - Calculate the output for each 'simple' component
 - Superimpose the outputs





Illustration







Clever Choice of the 'nominal input' a[t]

- To make the convolutions as simple as possible, it is best to chose a[t] to be Dirac Delta 'function'
- For any input function we can write

$$f_{in}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \delta[t - \tau] d\tau$$

The output is then just

$$f_{out}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \Delta[t - \tau] d\tau$$

where $\Delta[t]$ is the response of the circuit to a $\delta[t]$ input, the 'delta response':

$$\delta[t] \longrightarrow \text{in } H(s) \text{ out } \longrightarrow \Delta[t]$$

Note: I am a bit sloppy here with integration limits..





What is the Delta Response $\Delta[t]$?

• We do not know $\Delta[t]$, but: it turns out that its LT is just the transfer function!

The Laplace Transform of the Delta Response of a circuit is just given by its transfer function H[s]

■ Knowing that LT(∆[t]) = H[s], what is ∆[t] ? It's the Inverse LT:

$$\Delta[t] = LT^{-1} \{H[s]\}$$

- Why is this?
 - If we write down Kirchhoff's rules in the time domain, we get differential / integral equations.
 - The 'topology' of the equations is the same as using complex impedances.
 - If we transform this, we can get the impulse response





General Time Response

• Start from
$$f_{out}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \Delta[t - \tau] d\tau$$

Laplace transform both sides and use Convolution rule:

$$F_{\text{out}}[s] = LT \left\{ \int_{-\infty}^{\infty} f_{\text{in}}[\tau] \Delta[t - \tau] d\tau \right\} = F_{\text{in}}[s] LT \{\Delta[t]\}$$

■ Use our knowledge that LT {△[t]} = H[s]

$$F_{out}[s] = F_{in}[s] H[s]$$

And transform back:

$$f_{out}[t] = LT^{-1} \{ LT \{ f_{in}[t] \} H[s] \}$$

To calculate the time response of a circuit to an **arbitrary** input f[t]:

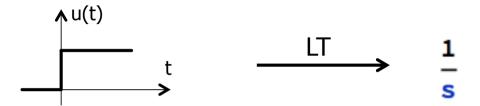
- 1. Laplace Transform f[t], yielding F[s]
- 2. Multiply with the Transfer function H[s]
- 3. Laplace Transform back



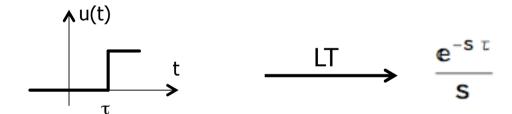


Important Input Functions

- The most important input to test a circuit is the Unit step:
 - It is often called u[t], Heaviside Step function, UnitStep,...



For a Shifted Step, use Time Shift rule:



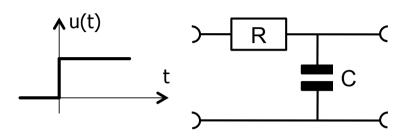
- A rectangular Pulse is just the difference of two Unit Steps
- For very short input signals (charge deposition in detector), input is the Dirac Delta, with LT = 1.





Example 1: Step Response of Low Pass

Consider



In[39]:=
$$H[s_] = \frac{1}{1 + s \tau}$$
;

▼ In[40]:= InverseLaplaceTransform[H[s], s, t]

Out[40]=
$$\frac{e^{-\frac{t}{\tau}}}{\tau}$$

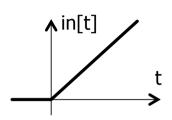
We knew that already...

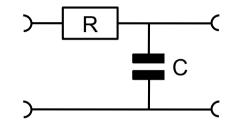




Example 2: Response of Low Pass to Slope

Now Consider a linear input ramp in[t] = k t





- The LT is
 Out[301]= $\frac{k}{s^2}$
- So our response is
 - f[t_] = InverseLaplaceTransform[IN[s] H[s], s, t]

