

TIME DOMAIN ANALYSIS (A VERY SUPERFICIAL APPROACH)

Laplace Transform (LT)

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- The Laplace Transform (LT) is an integral transform similar to the Fourier Transform
- The LT of a function f(t) is defined as

$$F(s) = \mathcal{L}\left\{f
ight\}(s) = \int_{0}^{\infty} f(t) e^{-st} \,\mathrm{d}t,$$

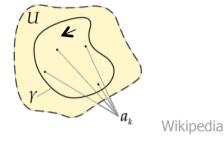
- s is a *complex* variable
- This integral does not necessarily exist for all possible f(t) and s (If s has a real part >0, f(t) must not grow faster than C e^{Re (s) t})
- The Inverse Transform is more complicated:

$$\mathcal{L}^{-1}\left\{F(s)
ight\}=rac{1}{2\pi i}\int_{\gamma-i\infty}^{\gamma+i\infty}e^{st}F(s)\,ds$$

- where γ > Re(all singularities of f).
- This is a *line integral* in the *complex s-plane*, 'right' of all singularities

The Residue Theorem states that the line integral of a function f(z) along a closed curve γ in the complex z-plane is 2πi × (the sum of the residues at the singularities a_k of f):

$$\oint_{\gamma} f(z) \, dz = 2\pi i \sum \operatorname{Res}(f, a_k)$$



- The residue is a characteristic of a singularity a_k (or c below)
 - For a first order (simple) *pole* at c (where f behaves ~ like 1/z):

$$\operatorname{Res}(f,c) = \lim_{z \to c} (z-c)f(z).$$

• More generally, for a pole of order n:

$$\operatorname{Res}(f,c) = \frac{1}{(n-1)!} \lim_{z \to c} \frac{d^{n-1}}{dz^{n-1}} \left((z-c)^n f(z) \right).$$

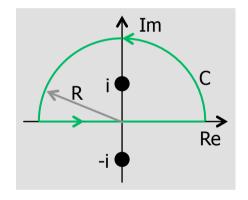
Example for Integration with Residues

- Assume we want to find $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1}$. The function $f(z) = \frac{1}{z^2 + 1} = \frac{1}{(z + i)(z i)}$ has poles i and -i
- The residue at the 'simple pole' i is:

$$Res(f,i) = \lim_{z \to i} f(z)(z-i) = \lim_{z \to i} \frac{1}{(z+i)} = \frac{1}{2z}$$

The line integral along green curve C is

$$\int_{C} f(z)dz = 2\pi i \ Res(f,i) = \pi$$



This is independent of R! With increasing R, the contribution of the upper arc vanishes (the length of the arc rises ~R, but f falls as 1/R²)

• Therefore
$$A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1} = \pi$$

Example 1: LT of 1

= For f(t) = 1:
L {1} =
$$\int_0^\infty e^{-st} dt = \left(\frac{-e^{-st}}{s}\right)_0^\infty = \frac{-e^{-s\infty}}{\sqrt{s}} + \frac{1}{\sqrt{s}}$$

Valid for Re[s] > 0 to make sure this vanishes

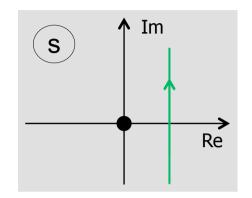
e^{-s 0}

S

S

Example 1 (inverse): Inverse LT of 1/s

For F(s) =
$$\frac{1}{s}$$
:
 $L^{-1}\left\{\frac{1}{s}\right\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} \frac{1}{s} ds$



- The integral has just one pole at s = 0.
- The Residuum is:

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$$\operatorname{Res}\left[e^{st}\frac{1}{s}, 0\right] = \operatorname{Limit}\left[(s-0)e^{st}\frac{1}{s}, s \to 0\right] = \operatorname{Limit}\left[e^{st}, s \to 0\right] = 1$$

So the Integral is

 $\int_{\gamma-i\infty}^{\gamma+i\infty} \dots ds = 2\pi i \operatorname{Res}[\dots, 0] = 2\pi i$

• And we just have
$$L^{-1}\left\{\frac{1}{s}\right\} = 1$$

(the arc contribution at infinity vanishes again...)

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Properties of Laplace Transforms

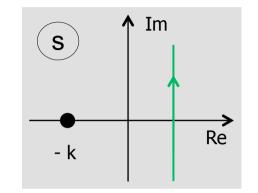
• For
$$f(t) = \mathcal{L}^{-1}{F(s)}, g(t) = \mathcal{L}^{-1}{G(s)}$$
 we have:

Function	Laplace Transform	
af(t)+bg(t)	aF(s)+bG(s)	Linearity
f'(t)	sF(s)-f(0)	Derivative
$(fst g)(t)=\int_0^t f(au)g(t- au)d au$	$F(s) \cdot G(s)$	Convolution
$\int_0^t f(\tau)d\tau = (u*f)(t)$	$rac{1}{s}F(s)$	Integration
f(t-a)u(t-a)u(t): Step function	$e^{-as}F(s)$	Time Shift

Example 2 ('frequency shift'):

$$\mathsf{L}^{-1}\left\{\mathsf{F}[\mathsf{s}]\right\} = \frac{1}{2 \pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{\mathsf{st}} \frac{1}{\mathsf{s}+\mathsf{k}} \, \mathrm{d}\mathsf{s}$$

• The pole is now at s = -k.



The Residuum is:

$$\operatorname{Res}\left[e^{st}\frac{1}{s+k}, 0\right] = \operatorname{Limit}\left[(s+k)e^{st}\frac{1}{s+k}, s \rightarrow -k\right] = e^{-kt}$$

• And we just have
$$L^{-1}\left\{\frac{1}{s+k}\right\} = e^{-kt}$$

Why is Laplace Transform so Useful ?

- Differential / Integral equations in t can be converted to Analytical equations in s, where they can be solved
- EQ(t) \rightarrow transform to H(s) \rightarrow Solve in s \rightarrow Transform back
- Example: Radioactive Decay
 - f[t]: Number of atoms at time t
 - The # of decaying atoms is prop. to # of atoms: $\frac{df[t]}{dt} = -\lambda f[t]$
 - With F[s] = LT(f[t]):
 (f[0] = N₀ is initial number of atoms)
 - This can be solved in s-domain:

$$s F[s] - f[0] = -\lambda F[s]$$

$$\int$$

$$F[s] = \frac{N_0}{s + \lambda}$$

• Transforming back (see example) gives: $f[t] = N_0 e^{-\lambda t}$

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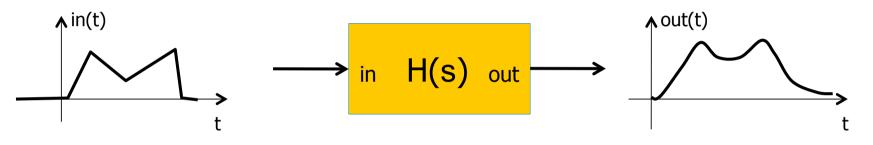


LAPLACE TRANSFORM AND TRANSFER FUNCTION

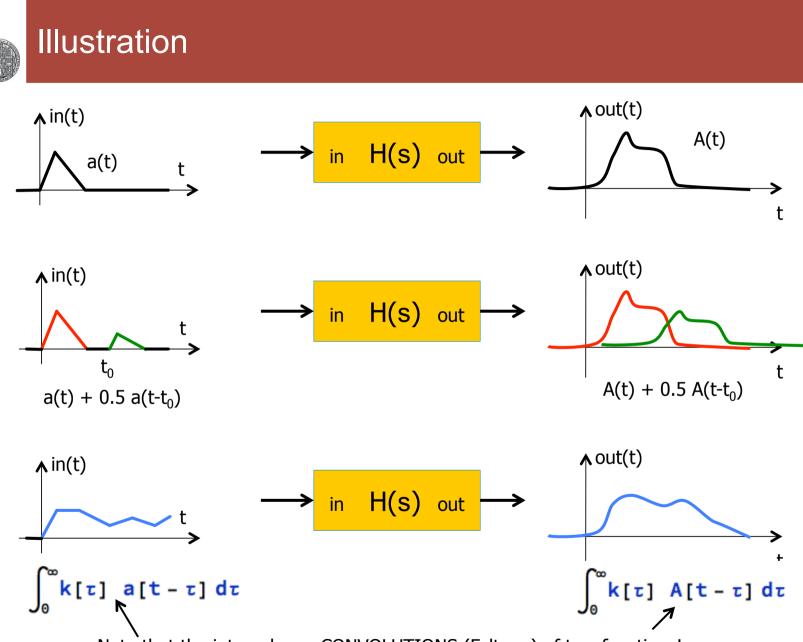


Time Response

- The transfer function tells us how sine inputs are modified by the system, i.e. what happen in the frequency domain
- How can be get the **time response** for an arbitrary input?



- For a *linear, time invariant (LTI)* system, we can use:
 - The response of a k × larger input pulse is just k × larger
 - The response for a time shifted input is time shifted
- For such a system we can
 - express the input signal as a superposition of 'simple' signals
 - Calculate the output for each 'simple' component
 - Superimpose the outputs



Note that the integrals are CONVOLUTIONS (Faltung) of two functions!

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Clever Choice of the 'nominal input' a[t]

- To make the convolutions as simple as possible, it is best to chose a[t] to be Dirac Delta 'function'
- For any input function we can write

$$f_{in}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \delta[t - \tau] d\tau$$

The output is then just

$$f_{out}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \Delta[t - \tau] d\tau$$

where Δ [t] is the response of the circuit to a δ [t] input, the *'delta response*':

$$\delta[t] \longrightarrow \text{in } H(s) \text{ out} \longrightarrow \Delta[t]$$

Note: I am a bit sloppy here with integration limits..

What is the Delta Response Δ [t] ?

■ We do not know ∆[t], but: it turns out that its LT is just the transfer function!

The Laplace Transform of the Delta Response of a circuit is just given by its transfer function H[s]

 Knowing that LT(Δ[t]) = H[s], what is Δ[t] ? It's the Inverse LT:

 $\Delta[t] = LT^{-1} \{H[s]\}$

- Why is this?
 - If we write down Kirchhoff's rules in the time domain, we get differential / integral equations.
 - The 'topology' of the equations is the same as using complex impedances.
 - If we transform this, we can get the impulse response

General Time Response

• Start from
$$\mathbf{f}_{out}[t] = \int_{-\infty}^{\infty} \mathbf{f}_{in}[\tau] \Delta[t - \tau] d\tau$$

Laplace transform both sides and use Convolution rule:

$$F_{out}[s] = LT \left\{ \int_{-\infty}^{\infty} f_{in}[\tau] \Delta[t - \tau] d\tau \right\} = F_{in}[s] LT \{\Delta[t]\}$$

• Use our knowledge that LT {A[t]} = H[s]

$$F_{out}[s] = F_{in}[s] H[s]$$

And transform back:

$$f_{out}[t] = LT^{-1} \{ LT \{ f_{in}[t] \} H[s] \}$$

To calculate the time response of a circuit to an **arbitrary** input f[t]:

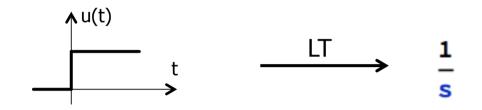
- 1. Laplace Transform f[t], yielding F[s]
- 2. Multiply with the Transfer function H[s]
- 3. Laplace Transform back

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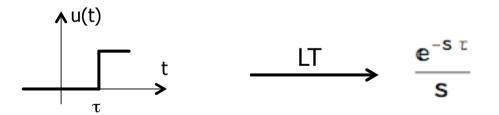


Important Input Functions

- The most important input to test a circuit is the Unit step:
 - It is often called u[t], Heaviside Step function, UnitStep,...



• For a Shifted Step, use Time Shift rule:

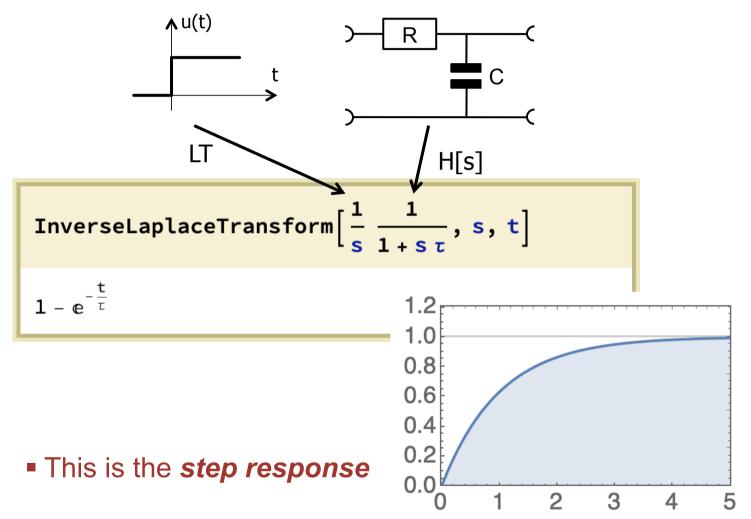


A rectangular Pulse is just the difference of two Unit Steps

For very short input signals (charge deposition in detector), input is the Dirac Delta, with LT = 1.

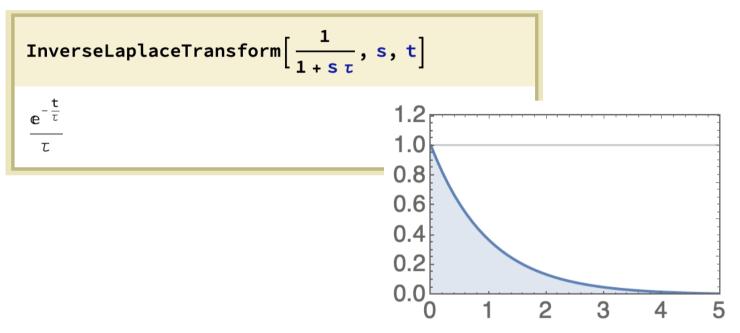
Example 1: Step Response of Low Pass

Consider





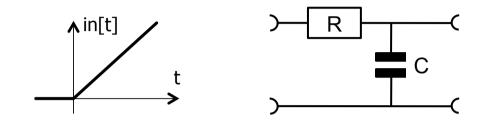
• What is LT⁻¹[H[s]] ???, e.g.



That is the output for a Dirac pulse at the input (with LT[δ[t]] = 1), i.e. that is the *impulse response*

Example 2: Response of Low Pass to Slope

Now Consider a linear input ramp in[t] = k t



r In[301]:= IN[s_] = LaplaceTransform[ktUnitStep[t], t, s]

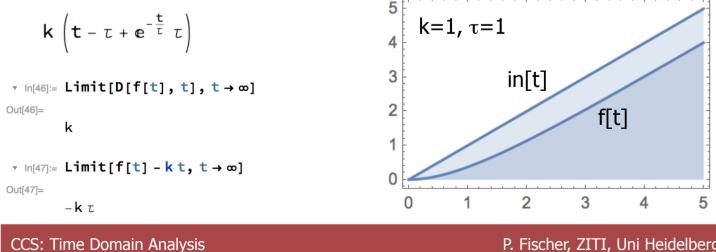


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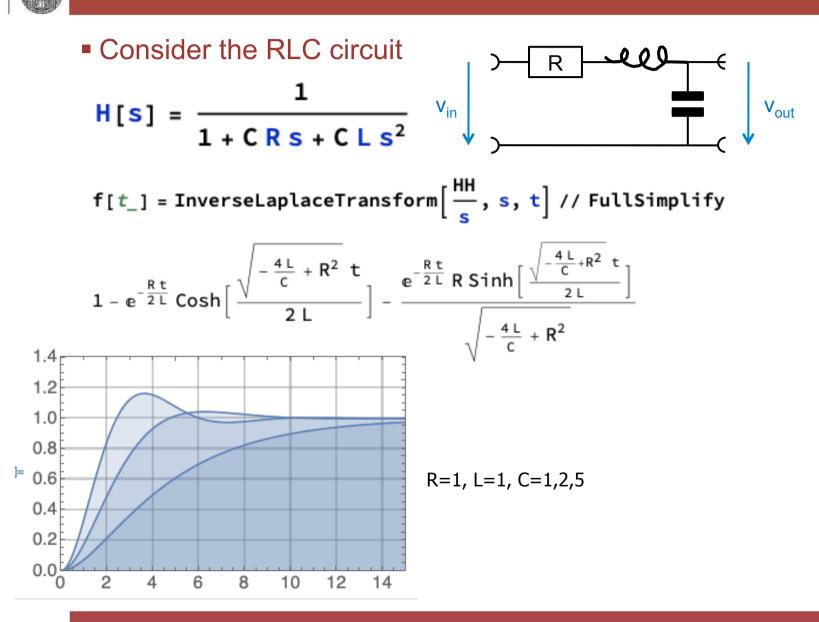
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So our response is

f[t_] = InverseLaplaceTransform[IN[s] H[s], s, t]



Example 3: Step Response of RLC



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