



**TIME DOMAIN ANALYSIS:
GET AN INTUITIVE UNDERSTANDING
(NO RIGOROUS DISCUSSION!)**



Laplace Transform (LT)

- The Laplace Transform (LT) is an integral transform similar to the Fourier Transform
- The LT of a function $f(t)$ is defined as

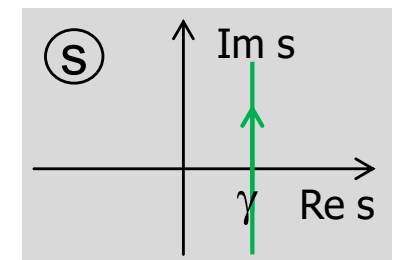
$$F(s) = \mathcal{L}\{f\}(s) = \int_0^{\infty} f(t)e^{-st} dt,$$

- s is a *complex* variable, so that $F(s)$ is also a complex value
- This integral does not necessarily exist for all possible $f(t)$ and s (If s has a real part >0 , $f(t)$ must not grow faster than $C e^{\text{Re}(s)t}$)

- The Inverse Transform is more complicated:

$$f(t) = \mathcal{L}^{-1}\{F(s)\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} F(s) ds$$

- where $\gamma > \text{Re}(\text{all singularities of } F)$.
- This is a *line integral* in the *complex s-plane*, 'right' of all singularities



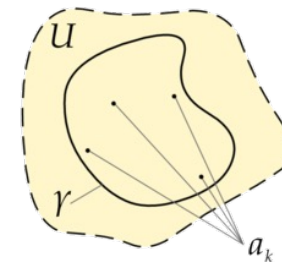


Reminder (hopefully..): Integration with Residues

This is very simplified!
 The statements are valid under certain conditions only.
 Consult a book on Complex Analysis!

- The *Residue Theorem* states that the line integral of a function $f(z)$ along a *closed* curve γ in the complex z -plane is $2\pi i \times$ (the sum of the residues at the *singularities* a_k of f):

$$\oint_{\gamma} f(z) dz = 2\pi i \sum \text{Res}(f, a_k)$$



Wikipedia

- The *residue* is a characteristic of a singularity a_k (or c below)
 - For a first order (simple) *pole* at c (where f diverges \sim like $1/z$):

$$\text{Res}(f, c) = \lim_{z \rightarrow c} (z - c) f(z).$$

- More generally, for a pole of order n :

$$\text{Res}(f, c) = \frac{1}{(n-1)!} \lim_{z \rightarrow c} \frac{d^{n-1}}{dz^{n-1}} ((z-c)^n f(z)).$$



Example for Integration with Residues

- Assume we want to find $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1} dx$.
- The function $f(z) = \frac{1}{z^2 + 1} = \frac{1}{(z + i)(z - i)}$ has poles i and $-i$

- The residue at the 'simple pole' i is:

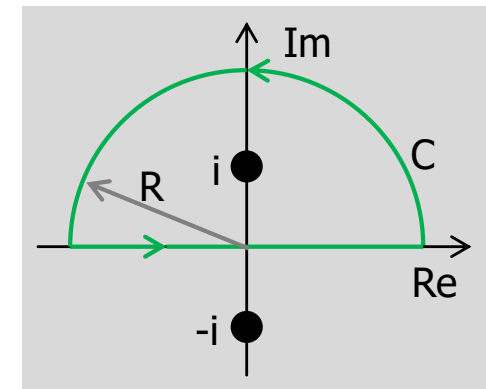
$$\text{Res}(f, i) = \lim_{z \rightarrow i} f(z)(z - i) = \lim_{z \rightarrow i} \frac{1}{(z + i)} = \frac{1}{2i}$$

- The line integral along green curve C is

$$\int_C f(z) dz = 2\pi i \text{Res}(f, i) = \pi$$

- This is independent of R! With increasing R, the contribution of the upper arc vanishes (the length of the arc rises $\sim R$, but f falls as $1/R^2$)

- Therefore $A = \int_{-\infty}^{\infty} \frac{1}{x^2 + 1} dx = \pi$





Example 1: LT of 1

- For $f(t) = 1$ (Unit step function):

$$L\{1\} = \int_0^{\infty} e^{-st} dt = \left(\frac{-e^{-st}}{s} \right)_0^{\infty} = \frac{-e^{-s\infty}}{s} + \frac{e^{-s0}}{s} = \frac{1}{s}$$

- Valid for $\text{Re}[s] > 0$ to make sure this vanishes



Example 1 (inverse): Inverse LT of 1/s

- For $F(s) = \frac{1}{s}$:

$$L^{-1} \left\{ \frac{1}{s} \right\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} \frac{1}{s} ds$$

- The integrand has just *one pole* at $s = 0$.
- It's a 1st order pole. Its Residuum is:

$$\text{Res} \left[e^{st} \frac{1}{s}, 0 \right] = \text{Limit} \left[(s - 0) e^{st} \frac{1}{s}, s \rightarrow 0 \right] = \text{Limit} [e^{st}, s \rightarrow 0] = 1$$

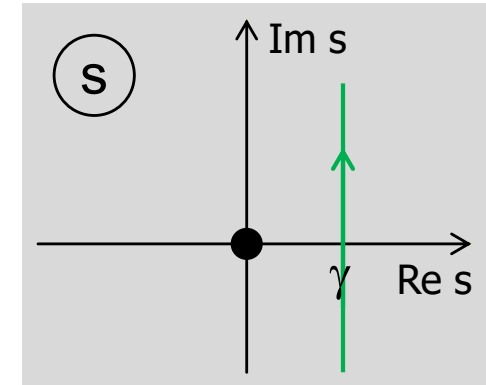
- So the Integral is (independent of γ , as long as $\text{Re}(\gamma) > 0$)

$$\int_{\gamma-i\infty}^{\gamma+i\infty} \dots ds = 2\pi i \text{Res} [\dots, 0] = 2\pi i$$

- And we just have

$$L^{-1} \left\{ \frac{1}{s} \right\} = 1$$

(the arc contribution at infinity vanishes again...)





Properties of Laplace Transforms

- For $f(t) = \mathcal{L}^{-1}\{F(s)\}$, $g(t) = \mathcal{L}^{-1}\{G(s)\}$ we have:

Function	Laplace Transform	
$af(t) + bg(t)$	$aF(s) + bG(s)$	Linearity
$f'(t)$	$sF(s) - f(0)$	Derivative
$(f * g)(t) = \int_0^t f(\tau)g(t - \tau) d\tau$	$F(s) \cdot G(s)$	Convolution ('Faltung')
$\int_0^t f(\tau) d\tau = (u * f)(t)$	$\frac{1}{s}F(s)$	Integration
$f(t - a)u(t - a)$	$e^{-as}F(s)$	Time Shift

u(t): Step function

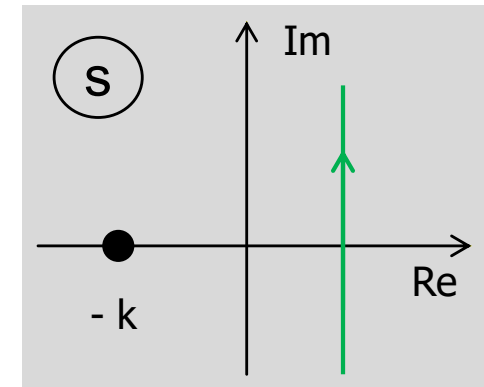


Example 2 ('frequency shift'):

- For $F(s) = \frac{1}{s+k}$:

$$L^{-1}\{F[s]\} = \frac{1}{2\pi i} \int_{\gamma-i\infty}^{\gamma+i\infty} e^{st} \frac{1}{s+k} ds$$

- The pole is now at $s = -k$.



- The Residuum is:

$$\text{Res}\left[e^{st} \frac{1}{s+k}, -k\right] = \text{Limit}\left[(s+k) e^{st} \frac{1}{s+k}, s \rightarrow -k\right] = e^{-kt}$$

- And we just have

$$L^{-1}\left\{\frac{1}{s+k}\right\} = e^{-kt}$$



Why is Laplace Transform so Useful ?

- Differential / Integral equations in t can be converted to analytical equations in s , where they can be solved
- EQ(t) \rightarrow transform to $H(s)$ \rightarrow Solve in s \rightarrow Transform back

- Example: Radioactive Decay

- $f[t]$: Number of atoms at time t

- The # of decaying atoms is prop. to # of atoms: $\frac{df[t]}{dt} = -\lambda f[t]$

- With $F[s] = \text{LT}(f[t])$:
($f[0] = N_0$ is initial number of atoms)

$$\text{LT} \curvearrowright s F[s] - f[0] = -\lambda F[s]$$

- This can be solved in s -domain:

$$F[s] = \frac{N_0}{s + \lambda}$$

- Transforming back (see example) gives: $f[t] = N_0 e^{-\lambda t}$

$$\text{LT}^{-1} \curvearrowright f[t] = N_0 e^{-\lambda t}$$

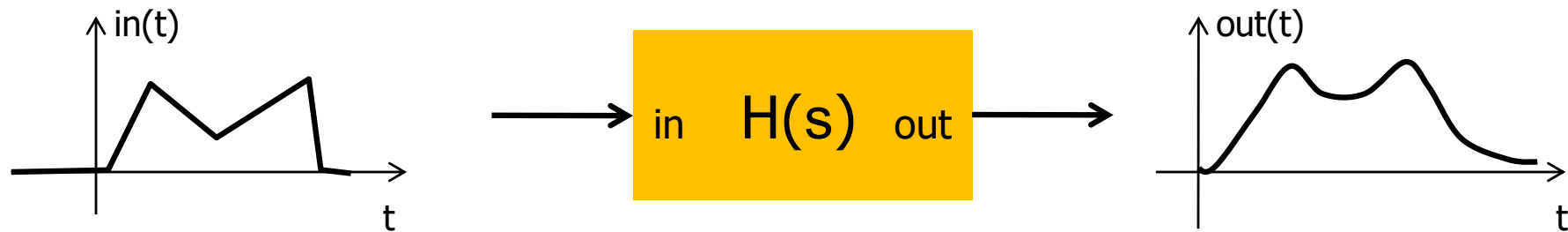


LAPLACE TRANSFORM AND TRANSFER FUNCTION



Time Response

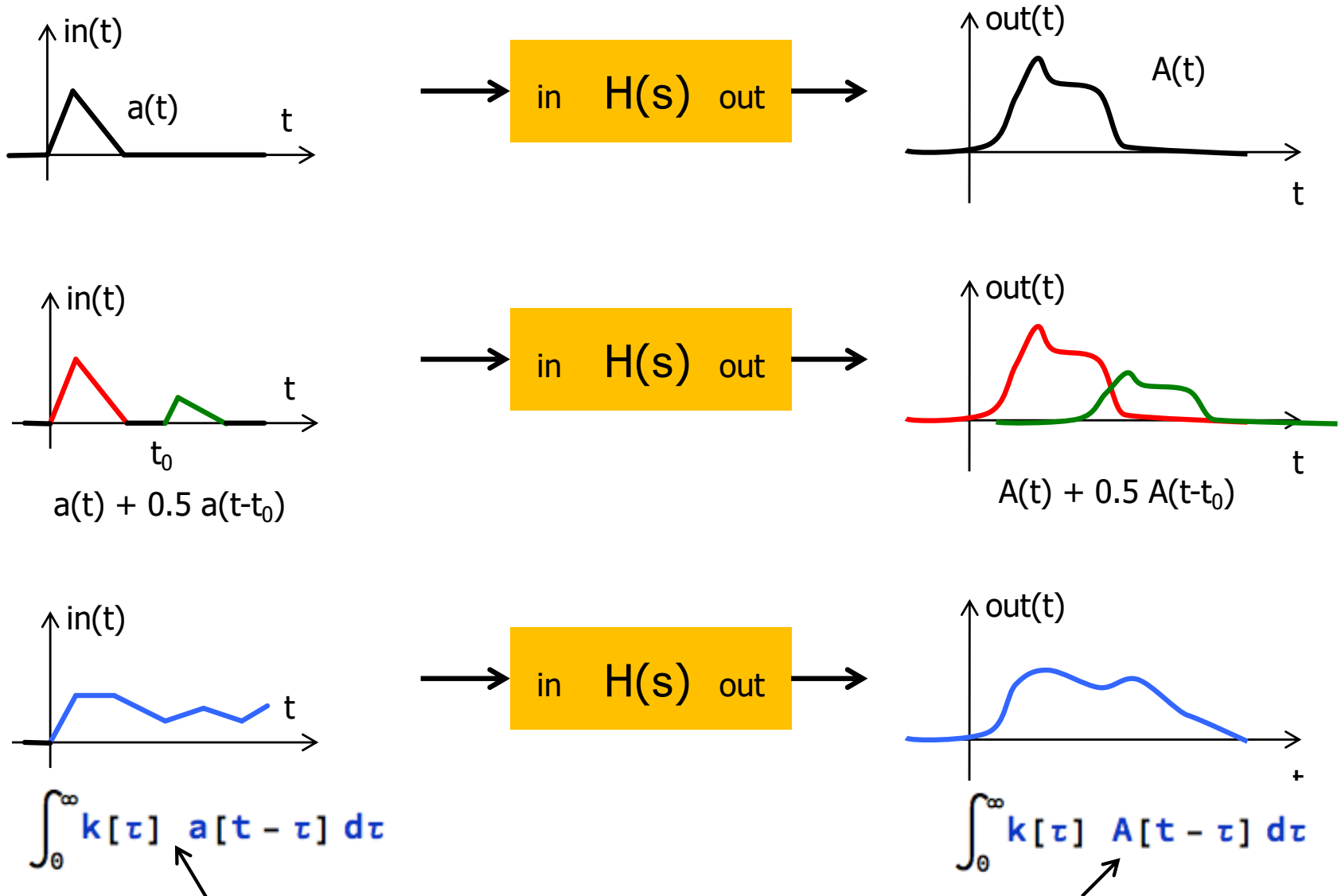
- The **transfer function** tells us how sine inputs are modified by the system, i.e. what happens in the **frequency domain**
- How can we get the **time response** for an arbitrary input?



- For a *linear, time invariant (LTI)* system, we can have:
 - The response of a $k \times$ larger input pulse is just $k \times$ larger
 - The response for a time shifted input is time shifted
- For such a system we can
 - express the input signal as a superposition of 'simple' signals
 - Calculate the output for each 'simple' component
 - Superimpose the outputs



Illustration



Note that the integrals are CONVOLUTIONS ('Faltung') of two functions!



Clever Choice of the 'nominal input' $a[t]$

- To make the convolutions as simple as possible, it is best to chose $a[t]$ to be Dirac Delta 'function'
- For *any* input function we can write

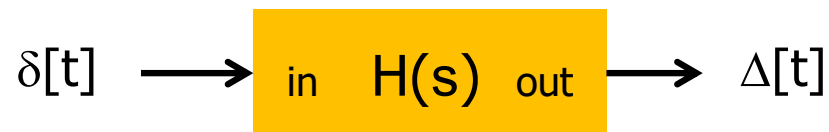
$$f_{in}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \delta[t - \tau] d\tau$$

!! The coefficients $k[\tau]$ on the previous page are simply f_{in} !

- The output is then just

$$f_{out}[t] = \int_{-\infty}^{\infty} f_{in}[\tau] \Delta[t - \tau] d\tau$$

where $\Delta[t]$ is the response of the circuit to a $\delta[t]$ input, the '**delta response**' or 'impulse response'



Note: I am a bit sloppy here with integration limits..



What is the Delta Response $\Delta[t]$?

- We do not know $\Delta[t]$, but: it turns out that its LT is just the transfer function!

The Laplace Transform of the Delta Response of a circuit is just given by its transfer function $H[s]$

- Knowing that $LT(\Delta[t]) = H[s]$, what is $\Delta[t]$?
It's the Inverse LT:

$$\Delta[t] = LT^{-1} \{H[s]\}$$

- Why is this?
 - If we write down Kirchhoff's rules in the time domain, we get differential / integral equations.
 - The 'topology' of the equations is the same as using complex impedances.
 - If we transform and solve them, we can get the impulse response



General Time Response

- Start from $f_{\text{out}}[t] = \int_{-\infty}^{\infty} f_{\text{in}}[\tau] \Delta[t - \tau] d\tau$

- Laplace transform both sides and use Convolution rule:

$$F_{\text{out}}[s] = \text{LT} \left\{ \int_{-\infty}^{\infty} f_{\text{in}}[\tau] \Delta[t - \tau] d\tau \right\} = F_{\text{in}}[s] \text{LT} \{ \Delta[t] \}$$

- Use our knowledge that $\text{LT} \{ \Delta[t] \} = H[s]$

$$F_{\text{out}}[s] = F_{\text{in}}[s] H[s]$$

- And transform back:

$$f_{\text{out}}[t] = \text{LT}^{-1} \{ \text{LT} \{ f_{\text{in}}[t] \} H[s] \}$$

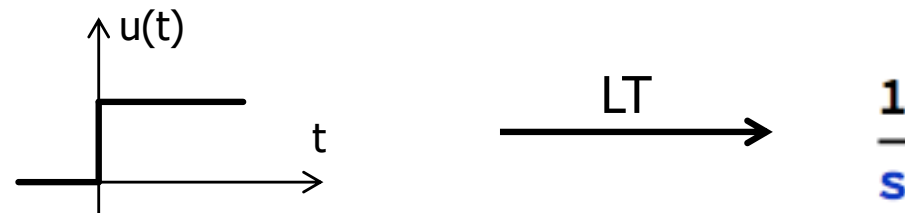
To calculate the time response of a circuit to an **arbitrary** input $f[t]$:

1. Laplace Transform $f[t]$, yielding $F[s]$
2. Multiply with the Transfer function $H[s]$
3. Laplace Transform back

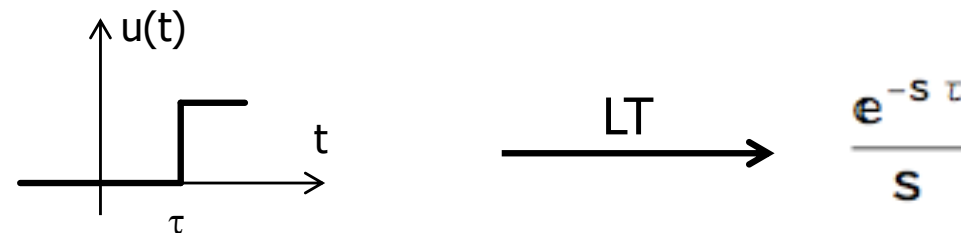


Important Input Functions

- The most important input to test a circuit is the Unit step:
 - It is often called $u[t]$, Heaviside Step function, UnitStep,...



- For a Shifted Step, use Time Shift rule:

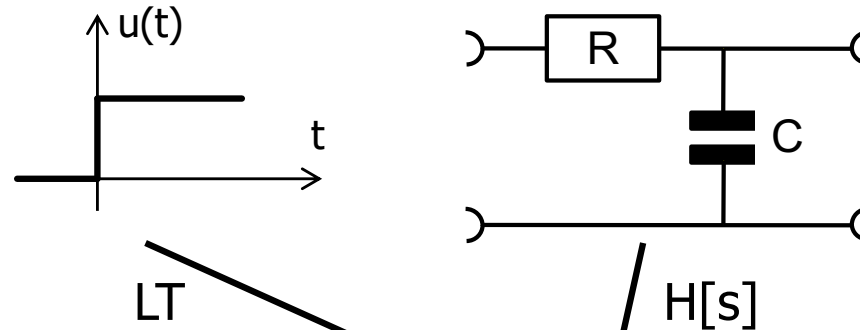


- A rectangular Pulse is just the difference of two Unit Steps
- For very short input signals (charge deposition in detector), input is the Dirac Delta, with $LT = 1$.



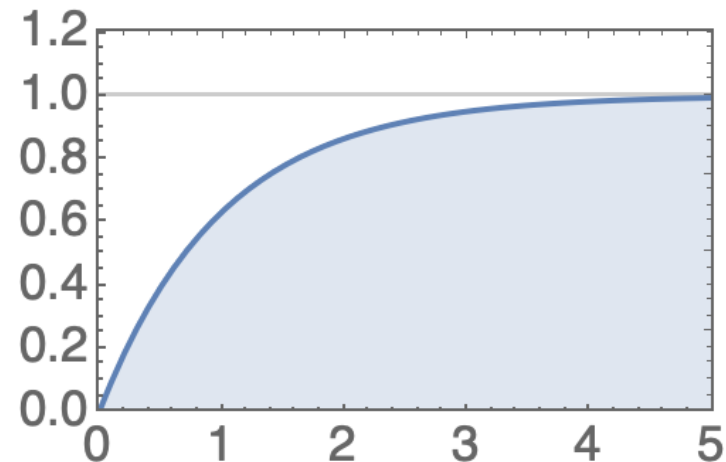
Example 1: Response of Low Pass to Step Input

- Consider



InverseLaplaceTransform $\left[\frac{1}{s} \frac{1}{1 + s\tau}, s, t \right]$

$$1 - e^{-\frac{t}{\tau}}$$



- This is the *step response*

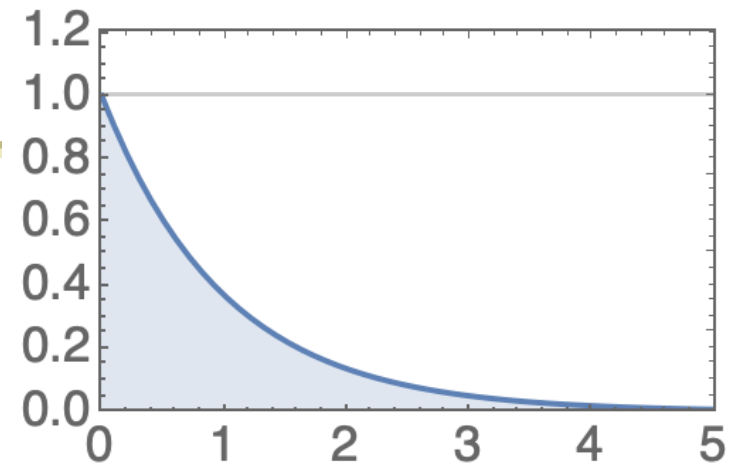


Watch out! Do not forget 1/s!

- What is $\text{LT}^{-1}[\text{H}[s]]$???, e.g.

$$\text{InverseLaplaceTransform}\left[\frac{1}{1 + s \tau}, s, t\right]$$

$$\frac{e^{-\frac{t}{\tau}}}{\tau}$$

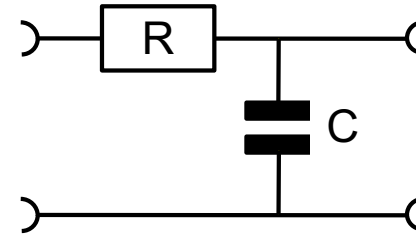
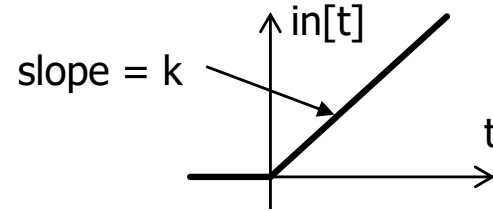


- That is the output for a **Dirac pulse** at the input (with $\text{LT}[\delta[t]] = 1$), i.e. that is the *impulse response*
- See page 14...



Example 2: Response of Low Pass to Slope

- Now Consider a linear input ramp : $in[t] = k t \text{ UnitStep}[t]$;



$$H[s] = \frac{1}{1 + s \tau}$$

- The LT is : $IN[t] = \text{LaplaceTransform}[in[t], t, s]$

$$\frac{k}{s^2}$$

Or using the definition $\int_0^{\infty} k t e^{s t} dt$

- The circuit response is

$$in[8] := out[t] = \text{InverseLaplaceTransform}[IN[s] H[s], s, t]$$

$$Out[8] = k \left(t + \left(-1 + e^{-\frac{t}{\tau}} \right) \tau \right)$$

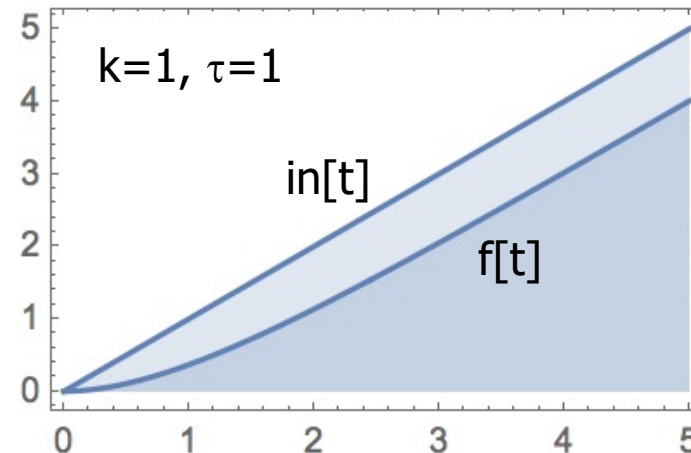
- Some Checks:

$$in[12] := \text{Limit}[D[out[t], t], t \rightarrow \infty]$$

$$Out[12] = k$$

$$in[11] := \text{Limit}[out[t] - in[t], t \rightarrow \infty]$$

$$Out[11] = -k \tau$$

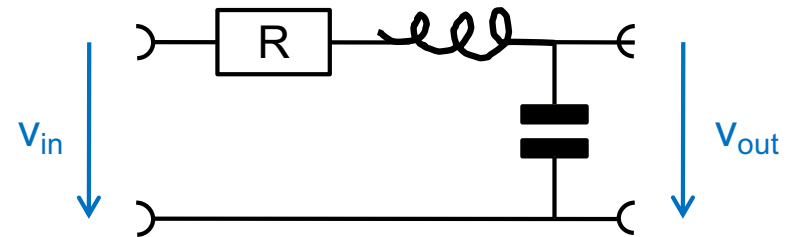




Example 3: Step Response of RLC

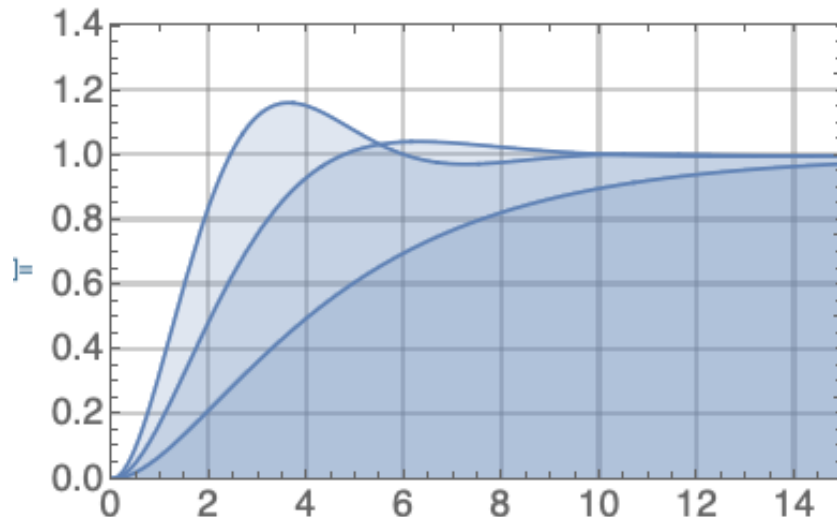
- Consider the RLC circuit

$$H[s] = \frac{1}{1 + CRs + CLs^2}$$



$$f[t_] = \text{InverseLaplaceTransform}\left[\frac{HH}{s}, s, t\right] // \text{FullSimplify}$$

$$1 - e^{-\frac{Rt}{2L}} \text{Cosh}\left[\frac{\sqrt{-\frac{4L}{C} + R^2} t}{2L}\right] - \frac{e^{-\frac{Rt}{2L}} R \text{Sinh}\left[\frac{\sqrt{-\frac{4L}{C} + R^2} t}{2L}\right]}{\sqrt{-\frac{4L}{C} + R^2}}$$



$$R=1, L=1, C=1, 2, 5$$